

# 1.0A Output, 2.5MHz Synchronous Buck-Boost DC/DC Converter

## FEATURES

- Regulated Output with Input Voltage Above, Below, or Equal to the Output.
- 1A Output Current at 3.3V in Step Down Mode ( $V_{IN} = 3.6V$  to  $5.5V$ ).
- Up to 800mA Output Current at 3.3V in Boost Mode ( $V_{IN} > 2.8V$ ).
- Single Inductor.
- 2.5V to 5.5V Input Voltage Range.
- Fixed and Adjustable Output Voltage Options from 1.8V to 5.5V.
- Up to 95% Efficiency.
- Stable with Low ESR Ceramic Capacitors.
- No Schottky Diode Required.
- Output Disconnect in Shutdown.
- $<1\mu A$  Shutdown Current.
- $<65\mu A$  Quiescent Current.
- Power Saving Mode for Improved Light Efficiency Operation.
- Forced Fixed Frequency Operation Mode.
- Load Disconnect During Shutdown.
- Undervoltage Lockout Protection.

## APPLICATIONS

- All Three-Cell Alkaline, NiCd or NiMH or Single-Cell Li Battery
- MP3 Players
- Handheld Instruments
- Digital Cameras
- Smart Phones
- Portable GPS Units
- Miniature Hard Disk Drive Power

## TYPICAL APPLICATION CIRCUIT

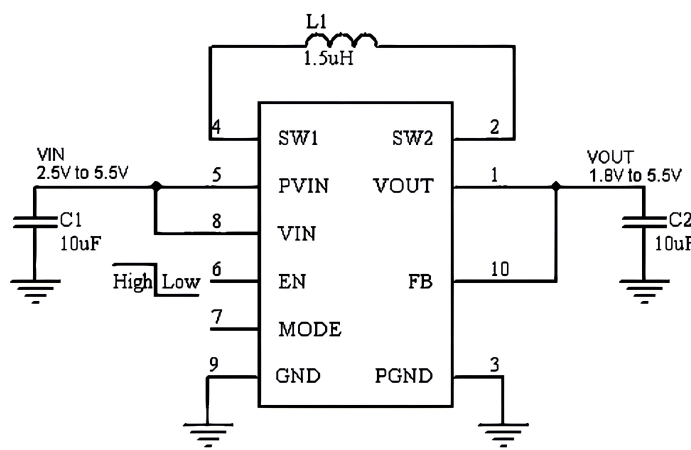


Fig.1 Application Circuit

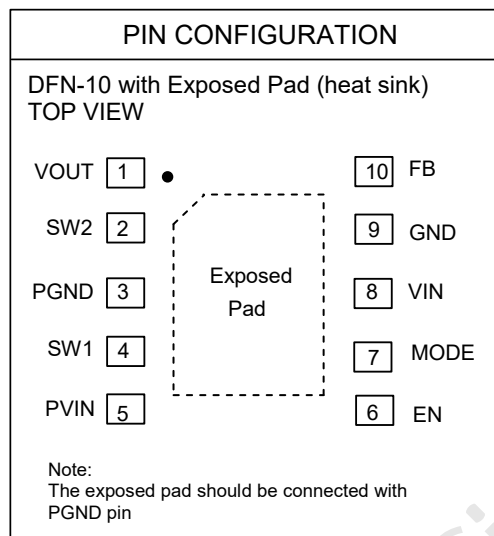
## DESCRIPTION

The PL9001 is a 1.0A output, low-noise, pulse-width-modulated (PWM) buck-boost DC-DC converter that operates from input voltage above, below, or equal to the output voltage.

The device features two internal synchronous rectifiers for high efficiency; it requires no external Schottky diode. Internally fixed-frequency 2.5MHz operation provides easy post-filtering and allows the use of small inductors and capacitors. At low load currents the converter enters the Power Saving Mode to maintain a high efficiency over a wide load range. The Power Saving Mode could be disabled, forcing PL9001 to operate at PWM mode. The PL9001 is ideally suited for single Li-Ion battery applications. It is also useful for three-cell alkaline, NiMH, or NiCd applications. Shutdown mode places the device in standby, reducing quiescent supply current to under  $1\mu A$ .

Other features of the PL9001 include internal soft-start, internal compensation, short circuit protection, current limit, and over temperature protection. The device is packaged in a 10-pin DFN package measuring 3 mm x 3 mm.

## ■ DEVICE INFORMATION



Order Information	Label Part NO.	Package	Package Qty	Top Marking
PL9001IDN10A	PL9001IDN10A	DFN3x3-10	5000	PL9001 RAAYMD

**PL9001:** Part Number

**RAAYMD:** RAA: LOT Code; YMD: Package Date Code

## ■ ABSOLUTE MAXIMUM RATINGS

Input Voltage Range on PVIN, VIN, SW1, SW2, VOUT, MODE, EN, FB .....	-0.3V to 6V
PGND to GND .....	-0.3V to 0.3V
Operating Ambient Temperature Range $T_A$ .....	-40°C~85°C
Operating Maximum Junction Temperature $T_J$ .....	150°C
Storage Temperature Range $T_{STG}$ .....	-65°C~150°C
Lead Temperature (Soldering 10 Sec.) .....	260°C
Thermal Resistance Junction to Case      DFN-10L 3x3 (with heat-sink)* .....	20°C/W
Thermal Resistance Junction to Ambient      DFN-10L 3x3 (with heat-sink)* .....	50°C/W
(Assume no Ambient Airflow)	

**Absolute Maximum Ratings are those values beyond which the life of a device may be impaired.**

\*The package is place on a two layers PCB with 2 ounces copper and 2 square inch, connected by 8 vias.

## ■ ELECTRICAL CHARACTERISTICS

( $V_{IN} = 3.6V$ ,  $V_{out} = 3.3V$ , unless otherwise specified. Typical values are at  $T_A = 25^{\circ}C$ ) (Note1)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
Input Voltage Range	$V_{IN}$		2.5		5.5	V
Output Adjustment Range	$V_{OUT}$		1.8		5.5	V
Feedback Voltage	$V_{FB}$		0.495	0.5	0.505	V
Feedback Input Impedance	$Z_{FB}$	$V_{OUT}=3.3V$		1.0		$M\Omega$
High Side Switch On Resistance	$R_{DSH\_ON}$			200		$m\Omega$
Low Side Switch On Resistance	$R_{DSL\_ON}$			200		$m\Omega$
Switch Current-Limit Threshold	$I_{sw}$			1.8		A
Quiescent Current	$I_Q$	$I_{OUT}=0mA$			65	$\mu A$
Shutdown Supply Current	$I_{SHDN}$	$V_{EN} = 0V$			1	$\mu A$
Oscillator Frequency	$f_{OSC}$		2.2	2.5	2.8	MHz
Line Regulation		$V_{IN}=2.5$ to $3.6V$ , $I_{OUT}=0$ to $500mA$ ,		0.5		%
Load Regulation		PWM mode		0.5		%
Under Voltage Lockout Threshold	$V_{UVLO\_R}$	$V_{IN}$ Rising	2.1		2.4	V
	$V_{UVLO\_F}$	$V_{IN}$ Falling	2.0		2.3	V
EN Low-Level Input Voltage	$V_{EN\_L}$				0.4	V
EN High-Level Input Voltage	$V_{EN\_H}$		1.2			V
Over Temperature Protection				150		$^{\circ}C$
Over Temperature Protection Hysteresis				30		$^{\circ}C$

Note 1: Specifications are production tested at  $T_A=25^{\circ}C$ . Specifications over the  $-40^{\circ}C$  to  $85^{\circ}C$  operating temperature range are assured by design, characterization and correlation with Statistical Quality Controls (SQC).

## ■ BLOCK DIAGRAMS

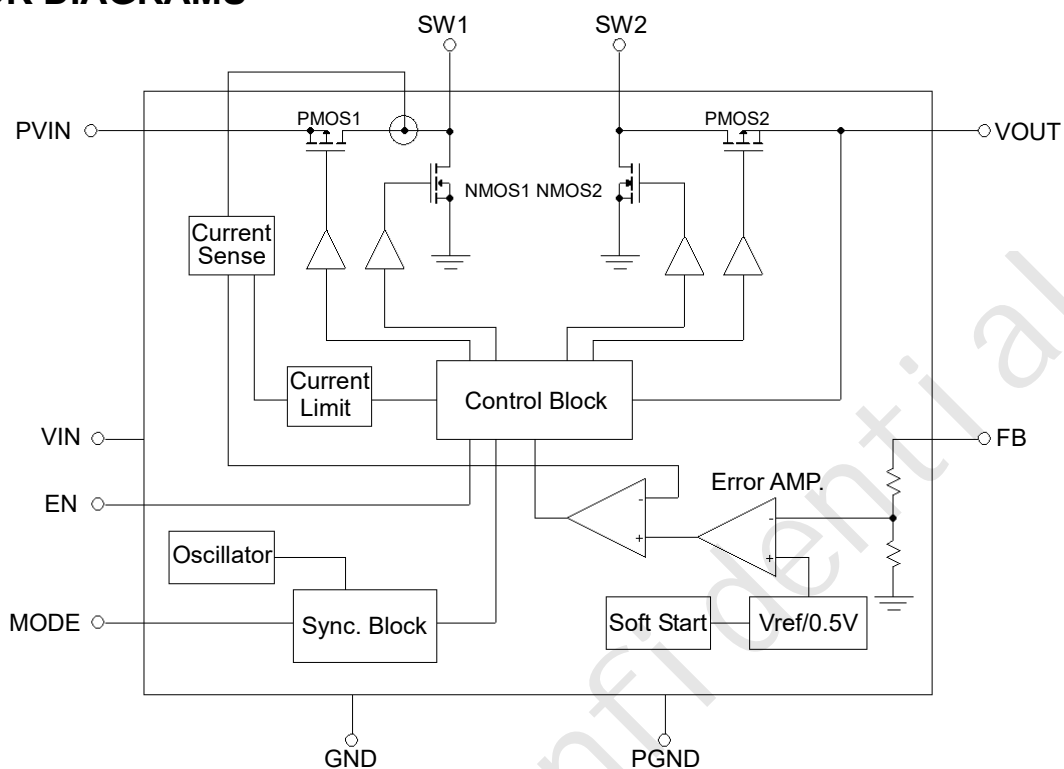


Fig.2 Fixed Output Voltage Version

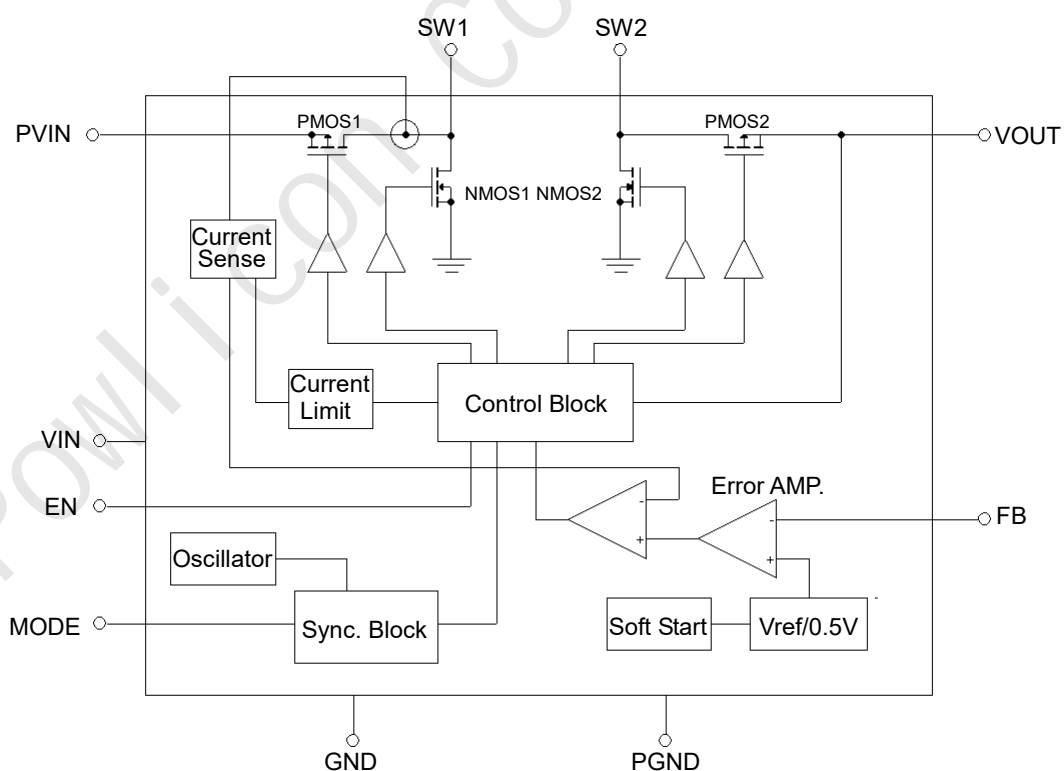


Fig.3 Adjustable Output Voltage Version

## PIN DESCRIPTIONS

PIN 1: VOUT – Buck-boost converter output.  
PIN 2: SW2 – Connection for inductor.  
PIN 3: PGND – Power ground.  
PIN 4: SW1 – Connection for inductor.  
PIN 5: PVIN – Supply voltage for power stage.  
PIN 6: EN – Enable input (1 enabled, 0 disabled).

PIN 7: MODE – Enable/Disable power save mode (1 disabled, 0 enabled, clock signal for synchronization).  
PIN 8: VIN – Supply voltage for control stage.  
PIN 9: GND – Control / logic ground.  
PIN 10: FB – Voltage feedback of adjustable version, must be connected to VOUT at fixed output voltage version.

## TYPICAL PERFORMANCE CHARACTERISTICS

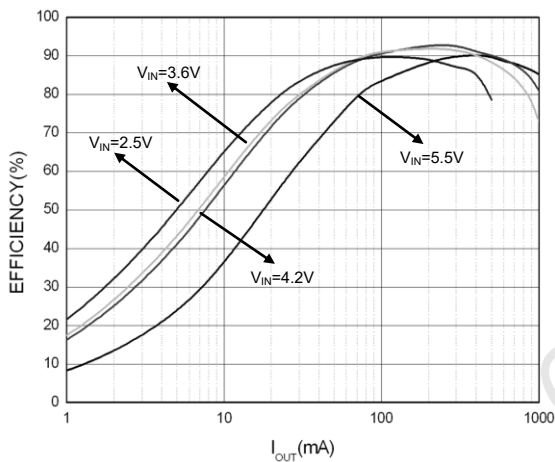


Fig.4 Efficiency vs. Output Current  $V_{OUT}=3.3V$  (PWM Mode)

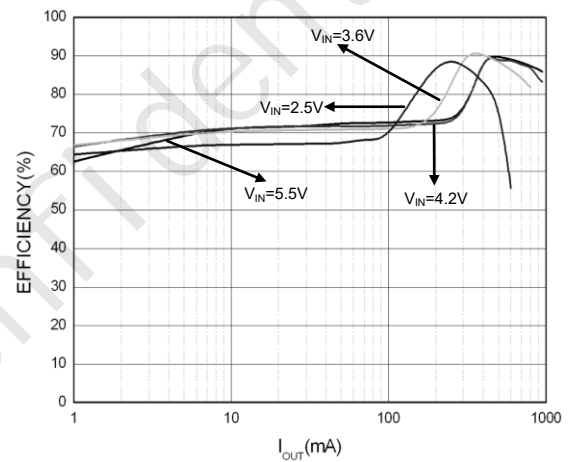


Fig.5 Efficiency vs. Output Current  $V_{OUT}=3.3V$  (PSM Mode)

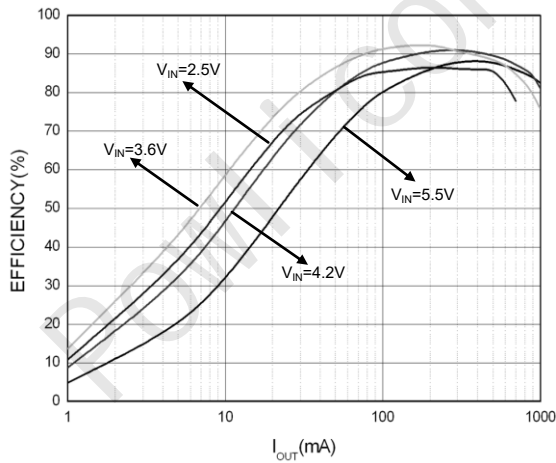


Fig.6 Efficiency vs. Output Current  $V_{OUT}=2.8V$  (PWM Mode)

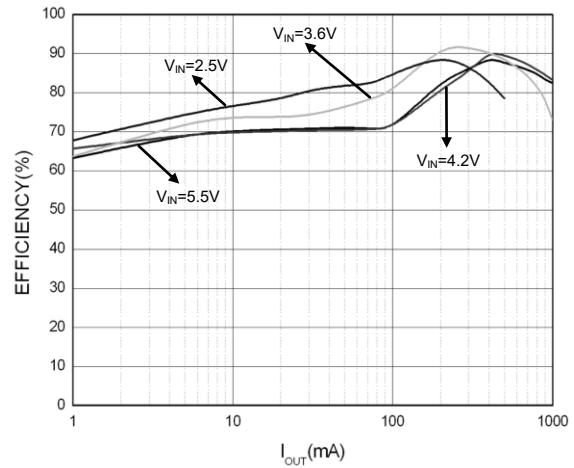


Fig.7 Efficiency vs. Output Current  $V_{OUT}=2.8V$  (PSM Mode)

## ■ TYPICAL PERFORMANCE CHARACTERISTICS (Continued)

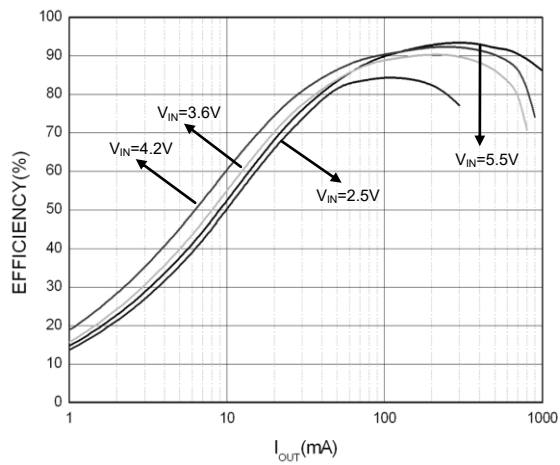


Fig.8 Efficiency vs. Output Current  $V_{OUT}=5V$  (PWM Mode)

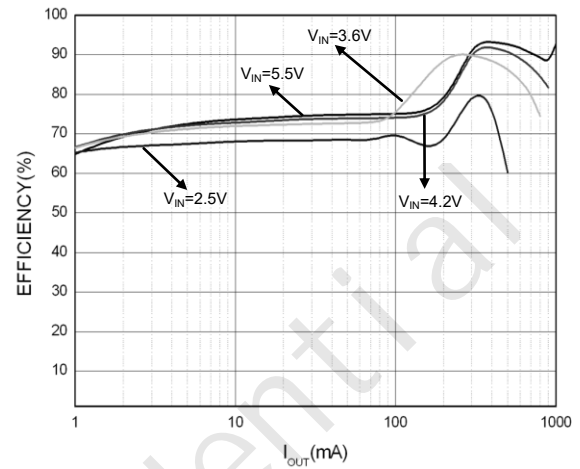


Fig.9 Efficiency vs. Output Current  $V_{OUT}=5V$  (PSM Mode)

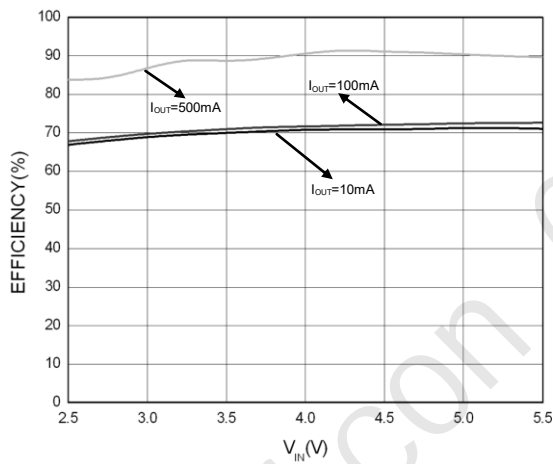


Fig.10 Efficiency vs. Input Voltage  $V_{OUT}=3.3V$  (PSM Mode)

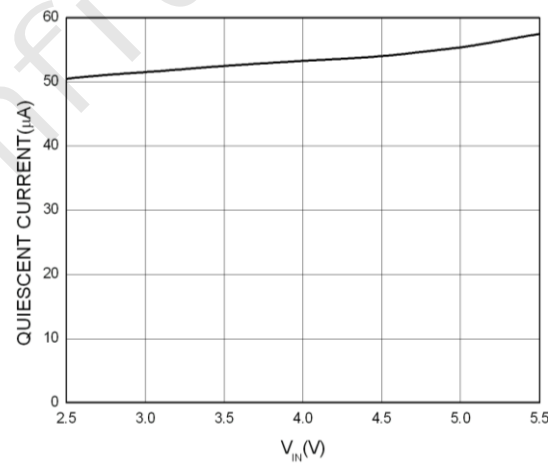


Fig.11 Quiescent Current vs. Input Voltage  $V_{OUT}=3.3V$

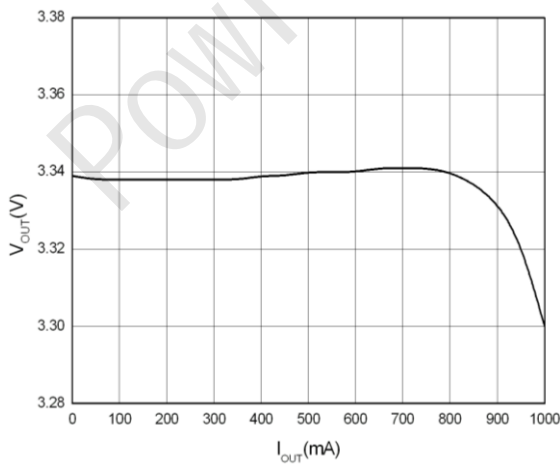


Fig.12 Load Regulation  $V_{IN}=3.6V$   $V_{OUT}=3.3V$  (PWM Mode)

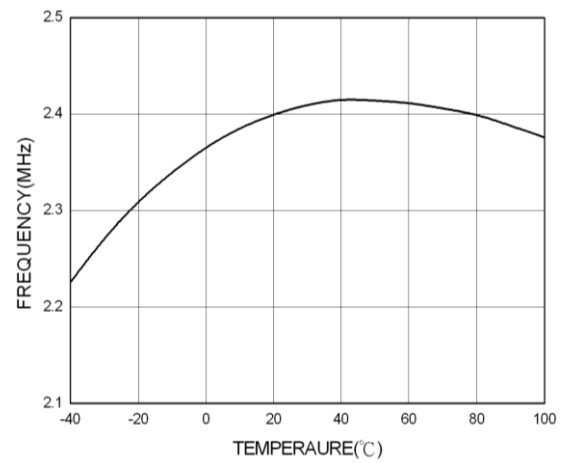


Fig.13 Frequency vs. Temperature ( $V_{IN}=3.6V$ )

## TYPICAL PERFORMANCE CHARACTERISTICS (Continued)

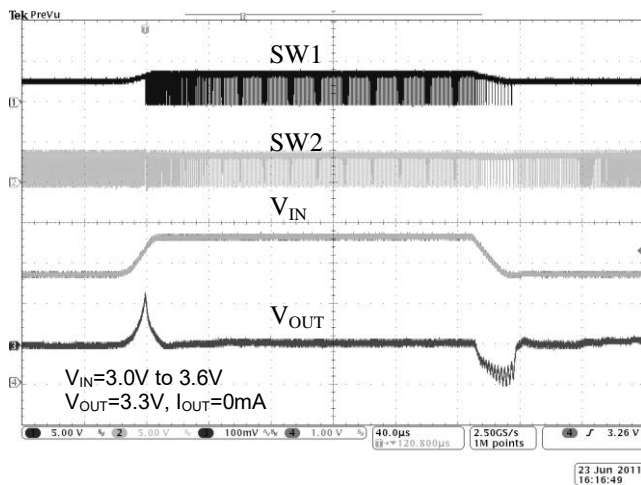


Fig.14 Line transient (PWM mode)

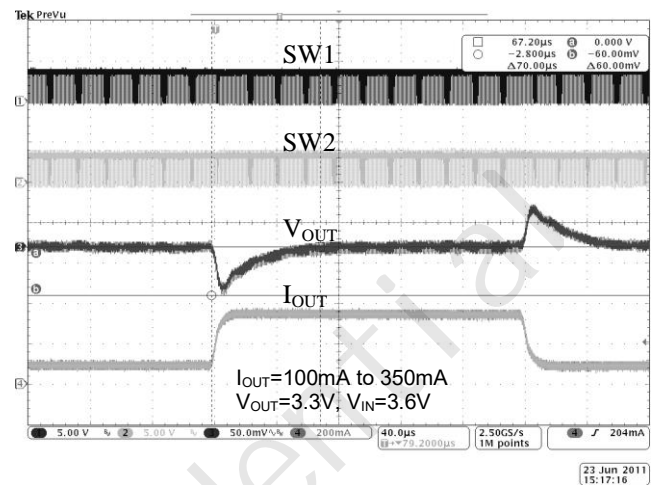


Fig.15 Load transient (PWM)

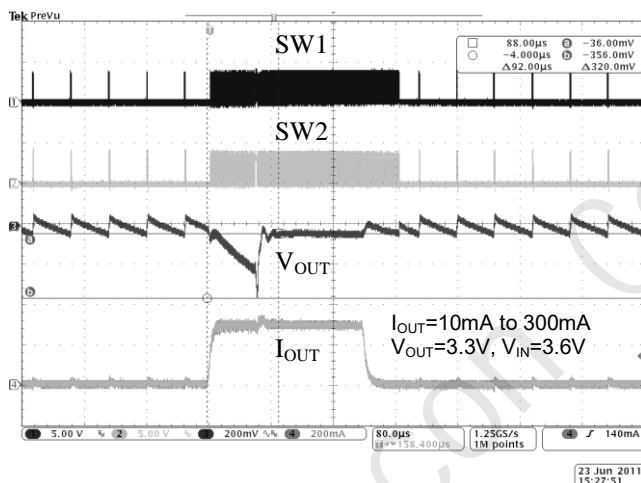


Fig.16 Load transient (PSM)

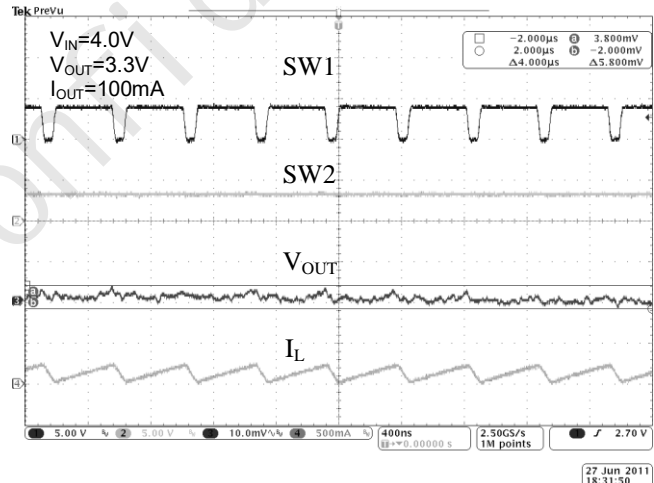


Fig.17. Buck operation

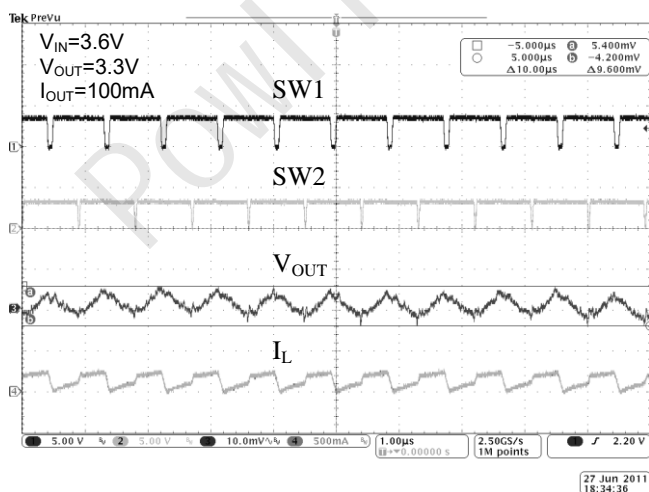


Fig.18 Buck-Boost Operation

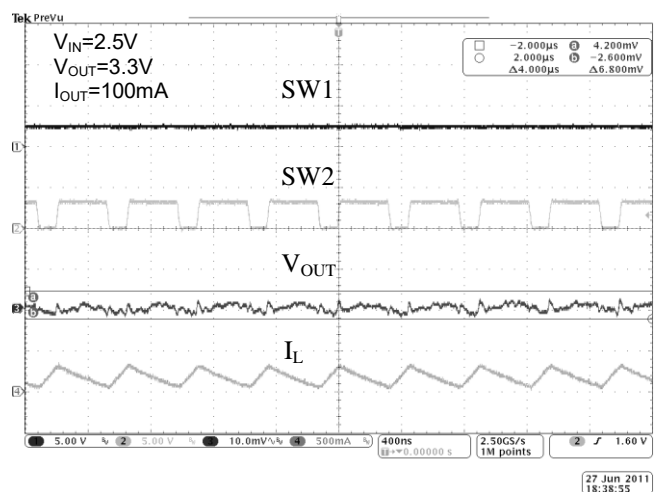


Fig.19 Boost Operation

## TYPICAL PERFORMANCE CHARACTERISTICS (Continued)

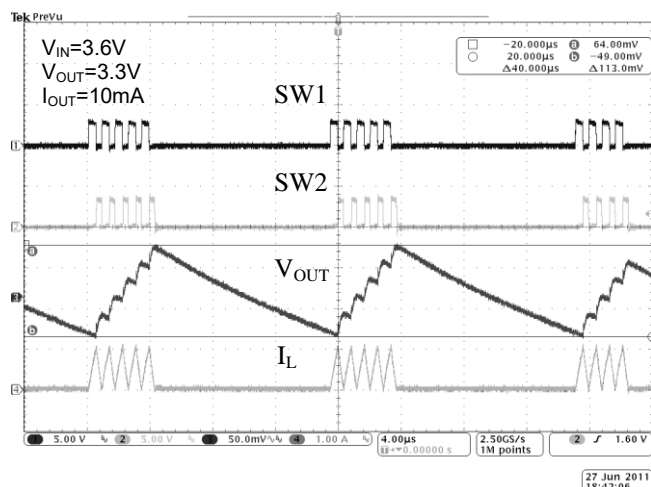


Fig.20 PSM Operation

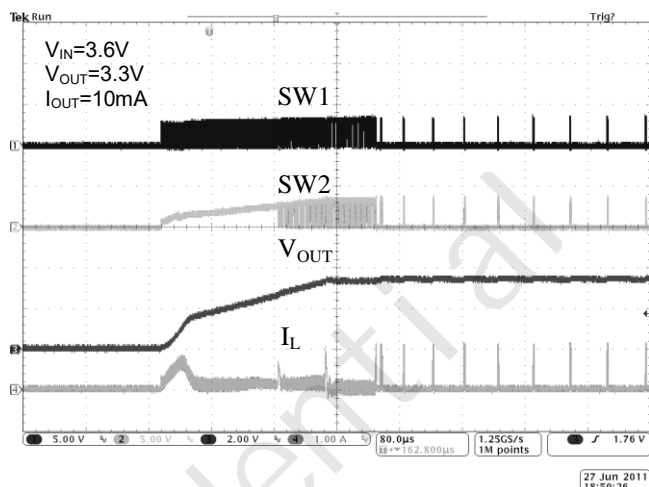


Fig.21 Power ON (PSM)

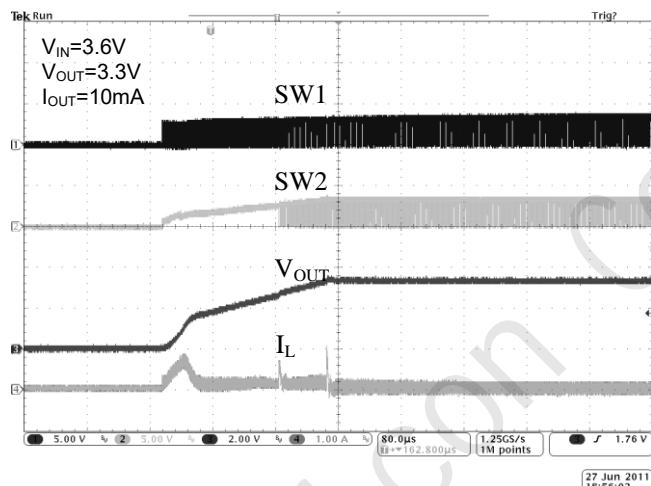


Fig.22 Power ON (PWM)

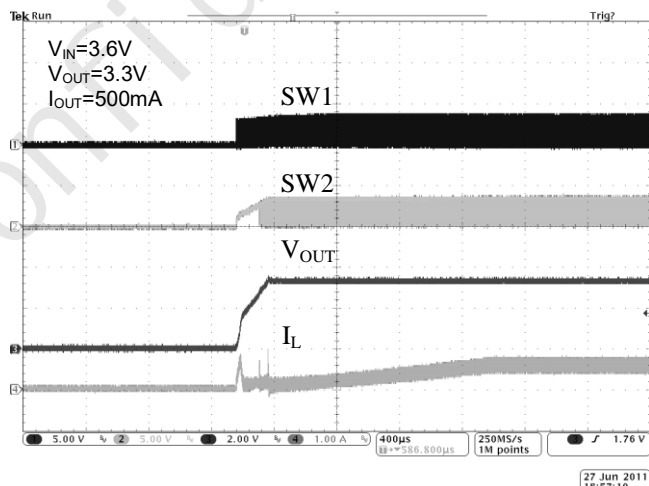


Fig.23 Power ON (PWM)

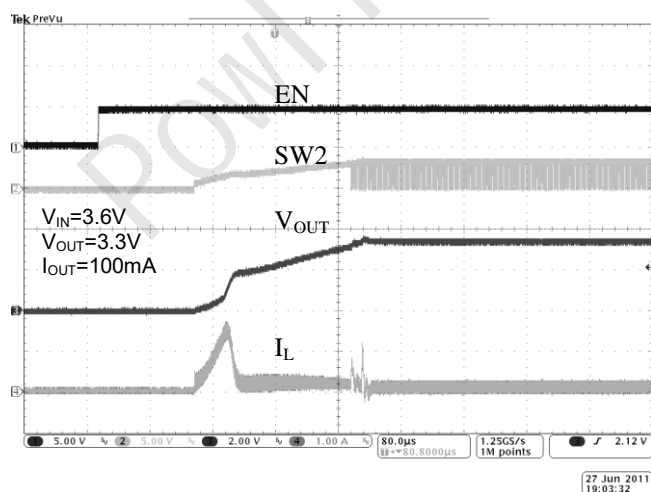


Fig.24 Start-up (PWM)

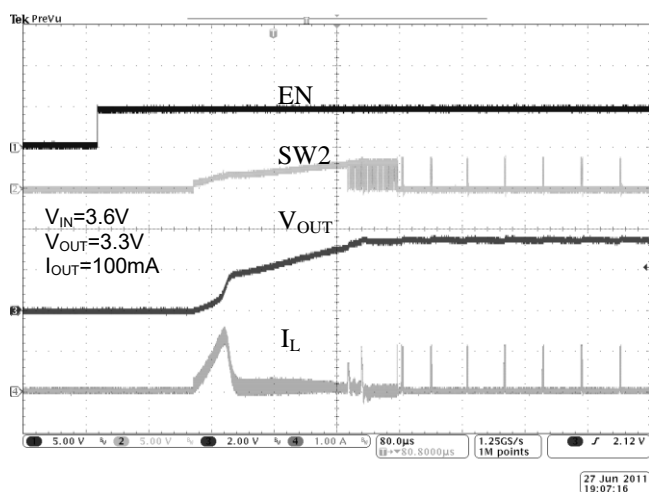


Fig.25 Start-up (PSM)

## ■ APPLICATION INFORMATION

### Operation

The buck-boost DC-DC converter operates in a synchronous current mode with a low-noise effect, which obtains the good regulation of output voltage from different input voltages. When the input voltage is higher than the output voltage, the synchronous switches PMOS1 and NMOS1 construct the step-down mode (Buck) while the PMOS2 is always turned on and the NMOS2 is always turned off. When the input voltage is lower than the output voltage, the synchronous switches PMOS2 and NMOS2 construct the step-up mode (Boost) while the PMOS1 is always turned on and the NMOS1 is always turned off. When the input voltage approaches the output voltage, the device can automatically operate at Buck-Boost mode.

### Enable

The device can be enabled when the voltage at EN pin is higher than 1.2V. It can go into the shutdown mode when the voltage at EN pin is lower than 0.4V. In shutdown mode, the PMOS1 and PMOS2 are switched off and the NMOS1 and NMOS2 are turned on.

### Power Saving Mode and Synchronization

The PL9001 can work at three different operation modes: the power saving mode, the forced fixed frequency mode and the external synchronization mode. When the voltage at MODE pin is low level, the PL9001 can enter the power saving mode (PSM) at light load condition. The PSM can improve the efficiency at light load condition. When the voltage at MODE pin is high level, the PL9001 can work at forced fixed frequency mode for all load conditions. Besides, by applying a 2.2MHz to 2.8MHz clock signal to the MODE pin, the PL9001 internal oscillator frequency can be synchronized to the external clock.

### Soft Start

The PL9001 provides the soft start function to avoid the overshoot of output voltage. During start-up period, an internal voltage ramp will clamp the

rising slope of output voltage. In order to avoid the inrush current during the start-up process, the output capacitance value should be considered.

### Current Limit

The PL9001 provides current limit function to limit the output power to protect the internal power switches. When the peak inductor current reaches 1.8A, the PMOS1 and the PMOS2 would be turned off.

### Inductor Selection

The component value of inductor and inductor current ripple are dependent. The calculation of inductance value uses the maximum input voltage in Buck mode and the minimum input voltage in Boost mode. The recommended minimum inductance value is greater than both of  $L(Buck)$  and  $L(Boost)$ .

$$L(Buck) \geq D \cdot \frac{(V_{IN\_MAX} - V_{OUT})}{f_{SW} \cdot \Delta I_L}, D = \frac{V_{OUT}}{V_{IN\_MAX}}$$

$$L(Boost) \geq D \cdot \frac{V_{IN\_MIN}}{f_{SW} \cdot \Delta I_L}, D = \frac{V_{OUT} - V_{IN\_MIN}}{V_{OUT}}$$

where the swithing frequency  $f_{SW}$  is 2.5MHz (Typically). The  $\Delta I_L$  is inductor current ripple and the  $D$  is duty cycle.

The efficiency of ferrite core is well. The saturation current of the inductor core must be higher than both of  $I_{peak}(Buck)$  and  $I_{peak}(Boost)$ . If the output current is fixed, the maximum peak current of inductor will appear while the PL9001 works at the boost mode.

$$I_{peak}(Buck) = 1.25 \cdot I_{OUT} + \frac{\Delta I_L}{2}$$

$$I_{peak}(Boost) = 1.25 \cdot I_{OUT} \cdot \frac{V_{OUT}}{V_{IN}} + \frac{\Delta I_L}{2}$$

### Capacitor Selection

The input capacitor should be put as close as possible to input terminal to avoid noise. The recommend input capacitor is the 10μF X5R/X7R

ceramic capacitor. For output capacitor, it's recommended to use low ESR ceramic capacitors. The output ripple voltage is given by

$$\Delta V_{OUT}(Buck) = D \cdot \frac{V_{IN\_MAX} - V_{OUT}}{8 \cdot L \cdot C_{OUT} \cdot f_{SW}^2}, D = \frac{V_{OUT}}{V_{IN\_MAX}}$$

$$\Delta V_{OUT}(Boost) = D \cdot \frac{I_{OUT}}{C_{OUT} \cdot f_{SW}}, D = \frac{V_{OUT} - V_{IN\_MIN}}{V_{OUT}}$$

### Output Voltage Programming

The PL9001 has two versions: one is fixed output voltage version, and another is adjustable version. The external resistors R1 and R2 are connected between output, FB and GND for the adjustable output voltage version. The reference voltage of the

feedback loop is 500mV. The recommended value of the feedback resistor R2 is 10kΩ. R1 is given by

$$R1 = R2 \cdot \left( \frac{V_{OUT}}{V_{FB}} - 1 \right)$$

### Layout Considerations

Placing the input capacitor, output capacitor and inductor as close to the PL9001 as possible and using short traces. The feedback resistor should be placed as close to the IC as possible. In order to get better performance, the wide and short trace should be used for large current loop.

## APPLICATION EXAMPLES

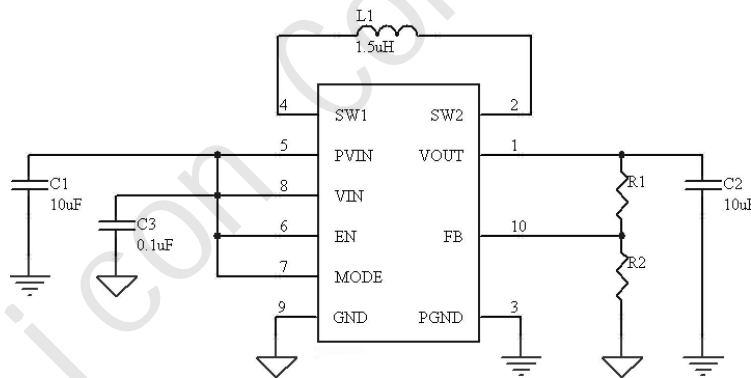
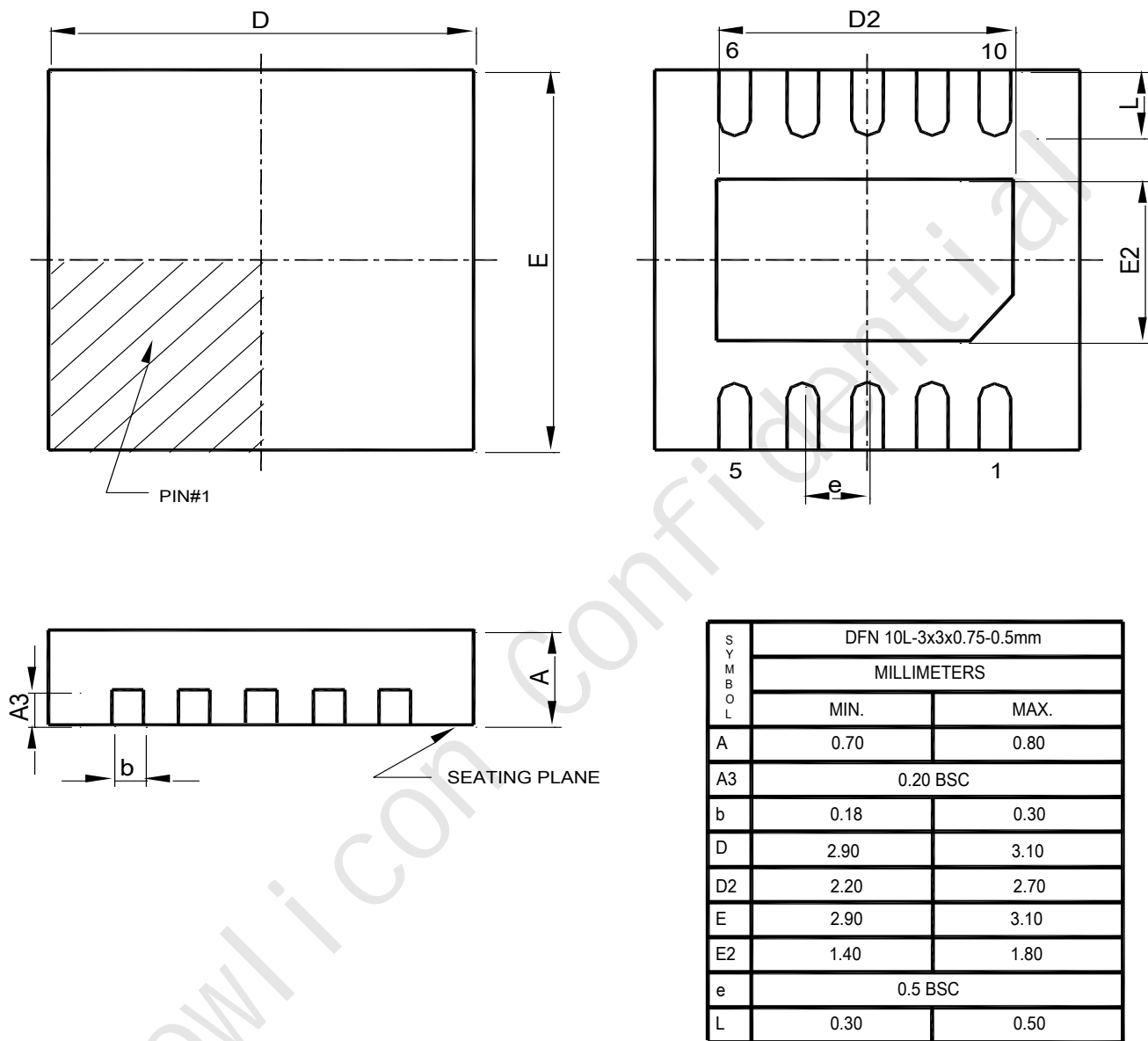


Fig.26 Adjustable Output Voltage

## ■ PHYSICAL DIMENSIONS (unit: mm)

### ● DFN-10L-3X3



Note : 1. DIMENSION AND TOLERANCING CONFORM TO ASME Y14.5M-1994.  
2. CONTROLLING DIMENSIONS : MILLIMETER , CONVERTED INCH DIMENSION ARE NOT NECESSARILY EXACT.

### Note:

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